

# Tuning rules for feedforward compensators combined with PID control



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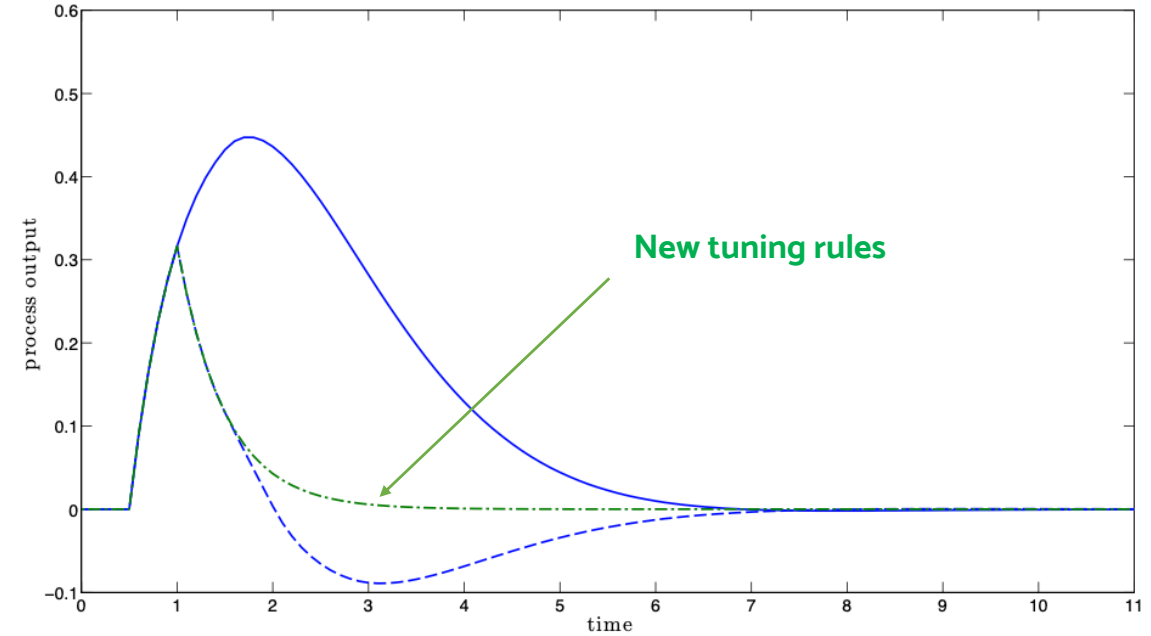
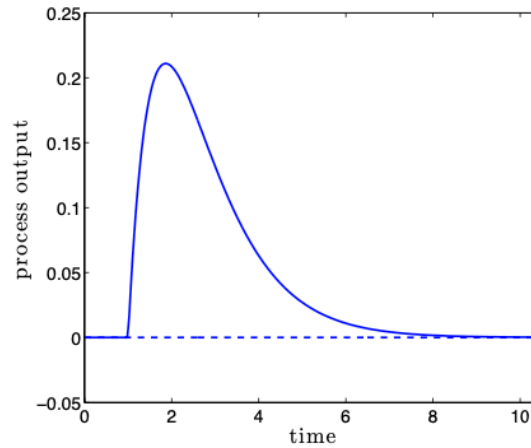
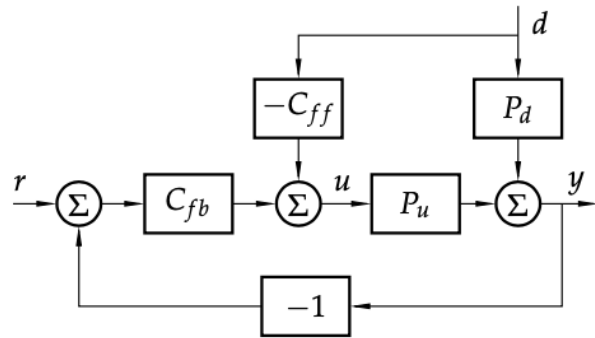
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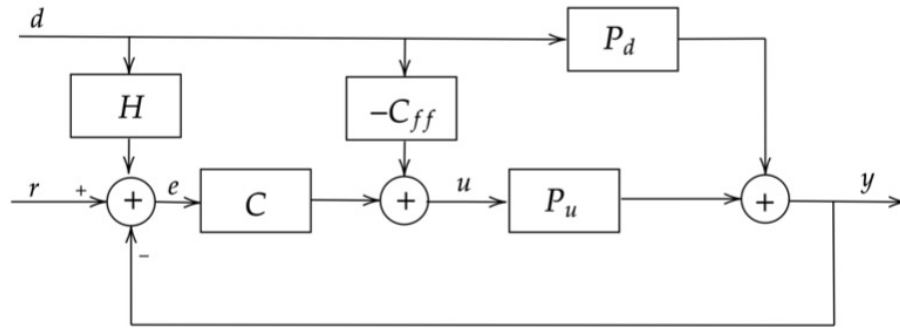
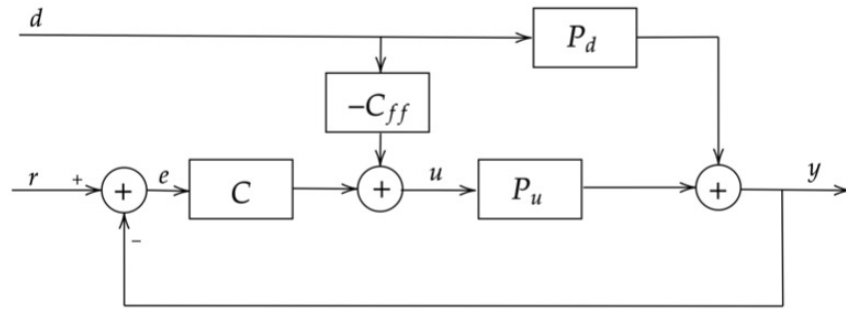
# What will we see in this presentation?



Ideal compensation:  $C_{ff} = \frac{P_d}{P_u} = P_d P_u^{-1}$

J. L. Guzmán, T. Hägglund. Tuning rules for feedforward control from measurable disturbances combined with PID control: A review. International Journal of Control, 2021.

# What will we see in this presentation?



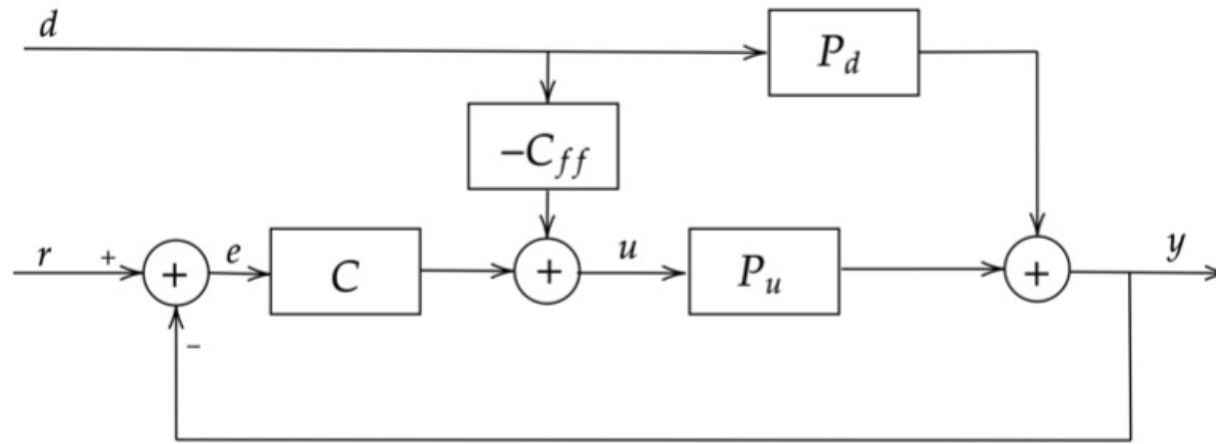
CS	Rule	M	$k_{ff}$	$T_p$	$T_z$
C	1	OS	$(K_d/K_u)e^{(L_d-L_u)/T_d}$	$T_d$	$T_u$
C	2	OS	$\frac{K_d(T_u+\lambda)}{K_u(T_d+\lambda)}\epsilon$ $T_u \neq T_d$ $\frac{K_d}{K_u}e^{(L_d-L_u)/T_d}$ $T_u = T_d$ $\epsilon = e^{(L_u/(T_u+\lambda)-L_d/(T_d+\lambda))}$	-	-
C	3	OS	$K_d/K_u - (K/\tau_i)IE$	$T_d$	$T_u$
C	4	OS	$K_d/K_u - (K/\tau_i)IE$	$T_d - (L_u - L_d)/4$	$T_u$
C	5	IAE	$(K_d(T_d + L_d))/(K_u(T_d + L_u))$	$T_d$	$T_u$
C	6	IAE	$\frac{K_d(T_u+\lambda)}{K_u(T_d+\lambda)}\theta$ $T_u \neq T_d$ $\frac{K_d(T_d+L_d)}{K_u(T_d+L_u)}$ $T_u = T_d$ $\theta = e^{-(L_u-L_d+(T_u-T_d)\log(2))/(T_d+\lambda)}$	-	-
C	7	IAE	$K_d/K_u - (K/\tau_i)IE$	$T_d - (L_u - L_d)/1.7$	$T_u$
C	8	ISE	$(K_d/K_u)e^{-(L_u-L_d)/(\lambda+T_d)}$	$T_d$	$T_u$
C	9	ISE	$K_d/K_u - (K/\tau_i)IE$	$T_d - (L_u - L_d)/\alpha$ $\alpha = \frac{L}{2T_d(1-e^{-L/(2T_d)})}$	$T_u$
C	10	IAE ISE OS	$K_d/K_u - (K/\tau_i)IE$	$T_d - (L_u - L_d)/\alpha$ $\alpha = \begin{cases} \frac{L}{2T_d(1-e^{-L/(2T_d)})} & \text{ISE} \\ 1.7 & \text{IAE} \\ 4 & \text{OS} \end{cases}$	$T_u$
B	11	ISE	$K_d/K_u$	$\frac{3a-1-b+(a-1)\sqrt{1+4b}}{b-2} T_d$ $b < 4a^2 - 2a$ or $0$ $b < a + \sqrt{a}$ otherwise $a = T_u/T_d$ $b = a(a+1)e^{L/T_d}$	$\eta = \left(1 - \frac{2T_u}{b(T_d+T_p)}\right)$
B	12	IAE ISE	$K_d/K_u$	$T_d - (L_u - L_d)/\alpha$ $\alpha = \begin{cases} \frac{L}{2T_d(1-e^{-L/(2T_d)})} & \text{ISE} \\ 1.7 & \text{IAE} \end{cases}$	$T_u$

$$IE = K_d(L_u - L_d + T_u - T_d + T_p - T_z)$$

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# 1

## Preliminaries



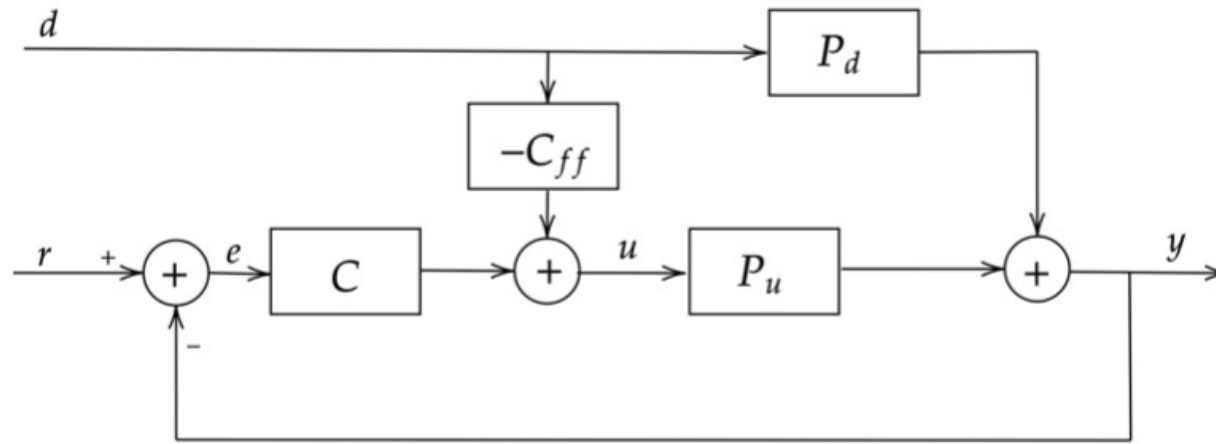
$$P_u = \frac{K_u}{1 + sT_u} e^{-sL_u}$$

$$P_d = \frac{K_d}{1 + sT_d} e^{-sL_d}$$

$$K = \frac{T_u}{K_u(\lambda + L_u)}, \quad \tau_i = T_u$$

$$C = K \left( 1 + \frac{1}{sT_i} \right)$$

$$\lambda > (0.5 + \sqrt{2})L_u$$



$$G_{y/d} = \frac{P_d - P_u C_{ff}}{1 + C P_u}$$

$$C_{ff} = k_{ff} \frac{sT_z + 1}{sT_p + 1} e^{-sL_{ff}}$$

$$C_{ff} = \frac{P_d}{P_u}$$

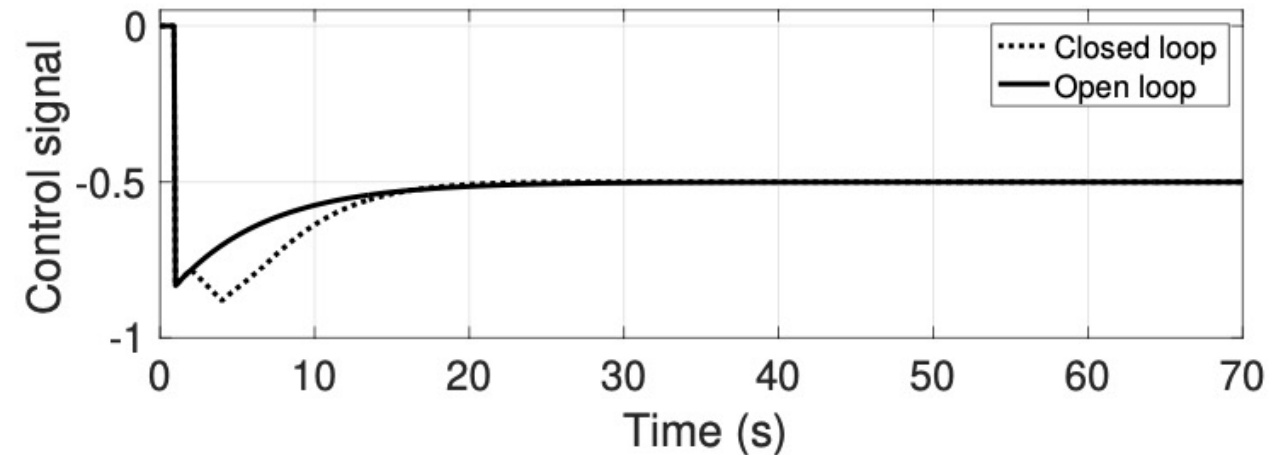
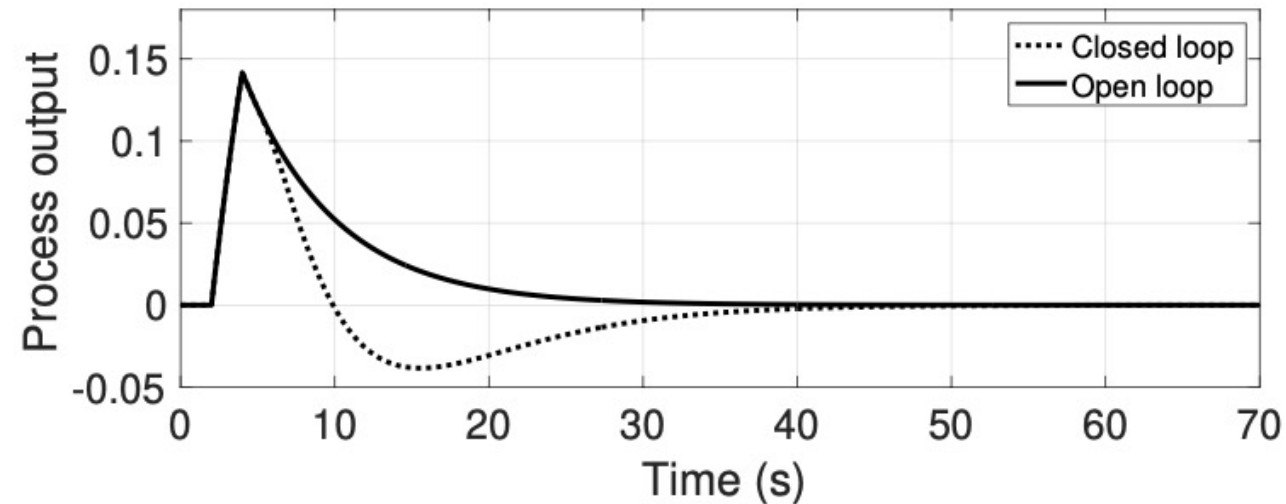
$$C_{ff} = \frac{K_d sT_u + 1}{K_u sT_d + 1} e^{-s(L_d - L_u)}$$

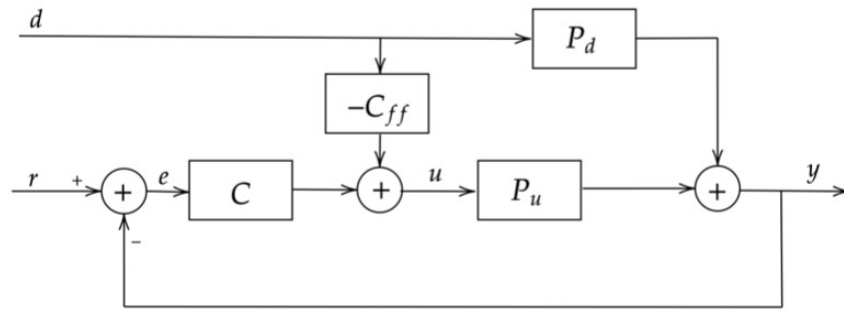
$$P_u = \frac{1}{1 + 10s} e^{-3s}, \quad P_d = \frac{0.5}{1 + 6s} e^{-s}$$

$$\tau_i = T_u = 10$$

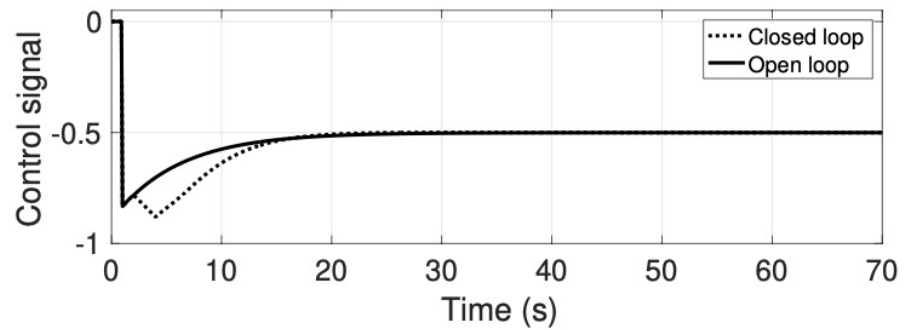
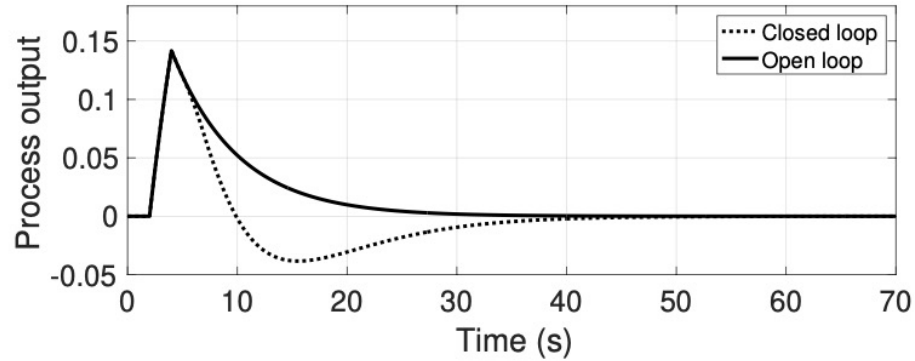
$$\lambda_2 = (0.5 + \sqrt{2})L_u$$

$$C_{ff} = \frac{K_d s T_u + 1}{K_u s T_d + 1} e^{-s(L_d - L_u)} = 0.5 \frac{10s + 1}{6s + 1}$$





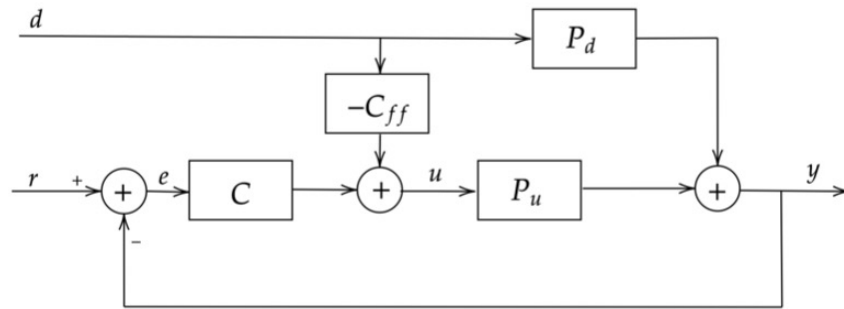
$$G_{y/d} = \frac{P_d - P_u C_{ff}}{1 + C P_u}$$



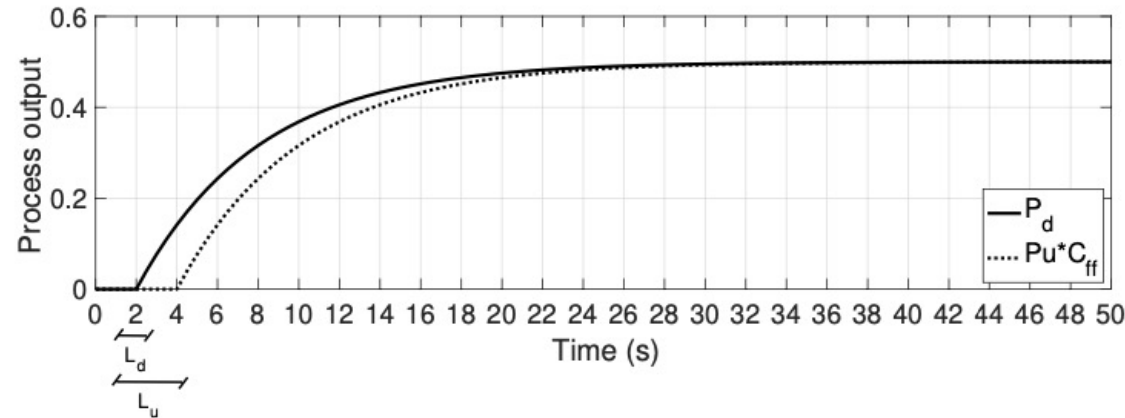
**Residual term!**

**Feedback and feedforward interaction!**

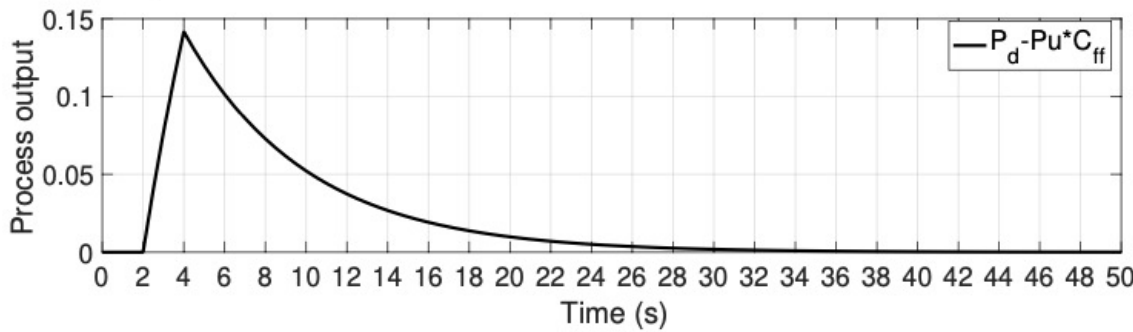




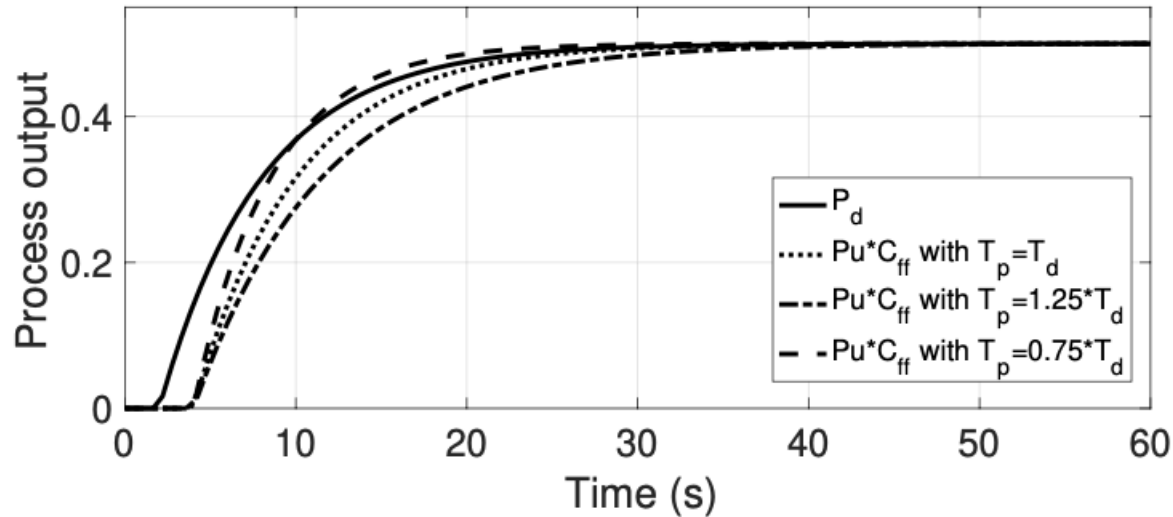
$$G_{y/d} = \frac{P_d - P_u C_{ff}}{1 + C P_u}$$



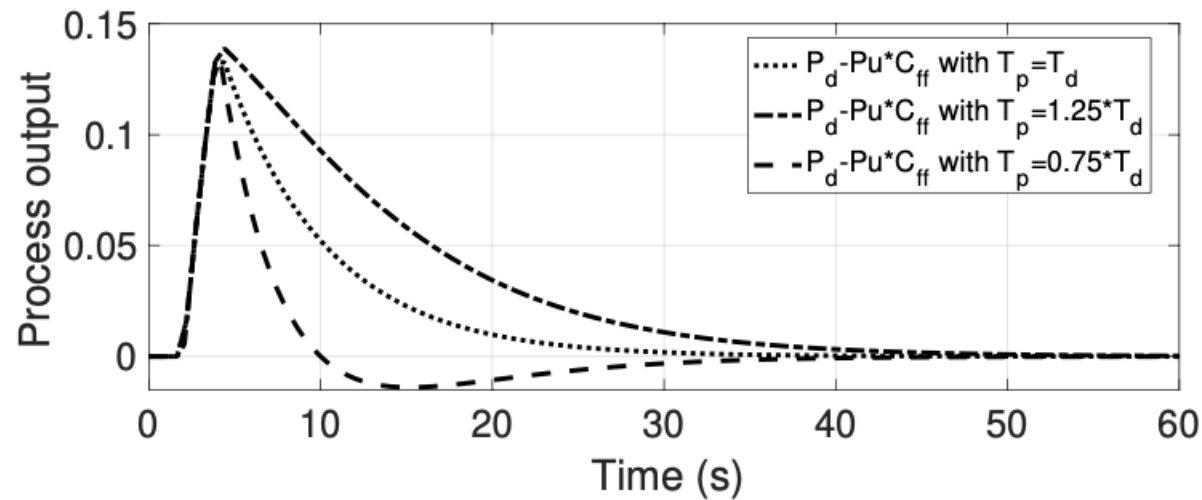
$$P_u = \frac{1}{1 + 10s} e^{-3s}, \quad P_d = \frac{0.5}{1 + 6s} e^{-s}$$



$$C_{ff} = \frac{K_d s T_u + 1}{K_u s T_d + 1} e^{-s(L_d - L_u)} = 0.5 \frac{10s + 1}{6s + 1}$$



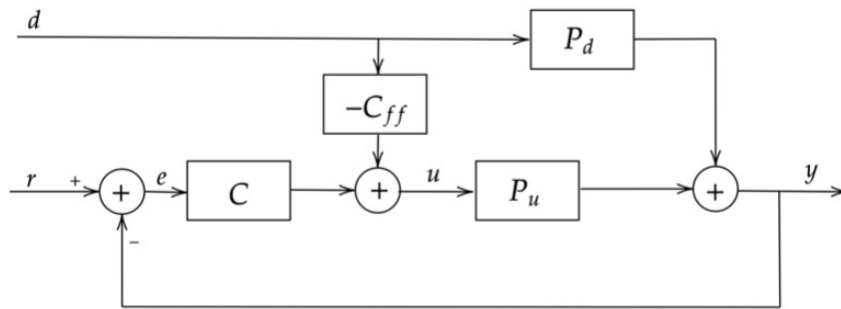
$$G_{ol} = P_d - P_u C_{ff}$$



**There is room for  
improvement!**

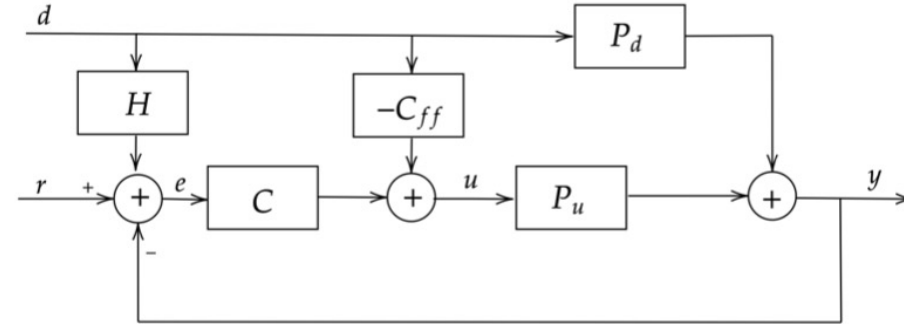
## Classical solutions for inversion problems in industry?

### Static feedforward



$$C_{ff} = k_{ff} = \frac{K_d}{K_u}$$

### Non-interactive scheme

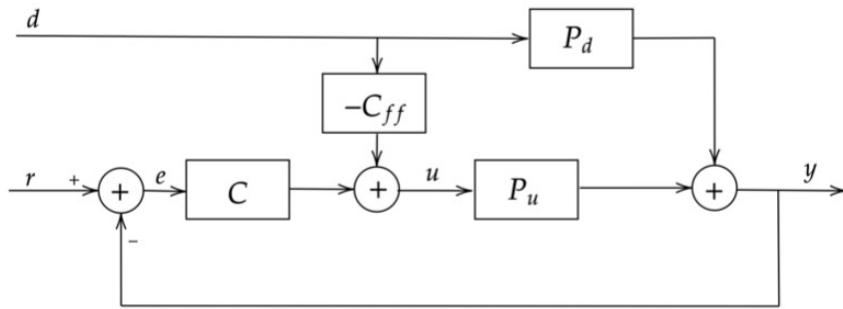


$$G_{e/d} = \frac{E(s)}{D(s)} = \frac{H - P_d + P_u C_{ff}}{1 + C P_u}$$

$$H = P_d - P_u C_{ff}$$

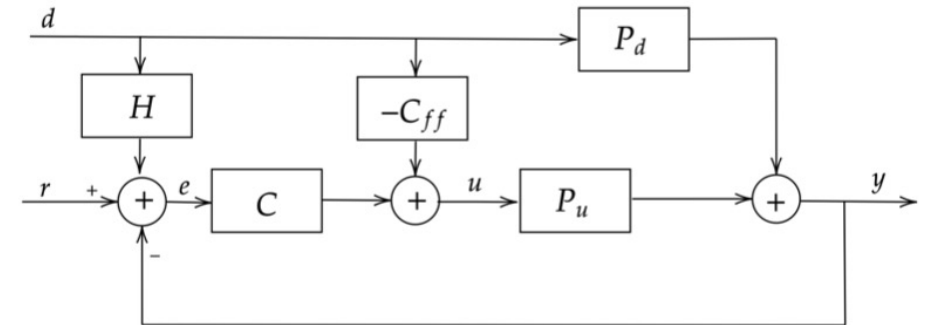
# 2 Tuning rules

## Classical scheme



Use the classic feedforward control scheme and tune the feedforward compensator properly. This means that the feedback controller  $C$  must be taken into account in the design.

## Non-interactive scheme



Use the non-interacting feedforward control scheme and tune the feedforward compensator properly. The design can be made without taking feedback controller  $C$  into account.

J. L. Guzmán, T. Häggglund. Tuning rules for feedforward control from measurable disturbances combined with PID control: A review. International Journal of Control, 2021.

## Inversion problems

- Non-realizable delay inversion.
- Right-half plane zeros.
- Integrating poles.

## Tuning rule objective

- Minimize IAE.
- Minimize ISE.
- Reduce overshoot.

**15 different tuning rules for feedforward compensators!**

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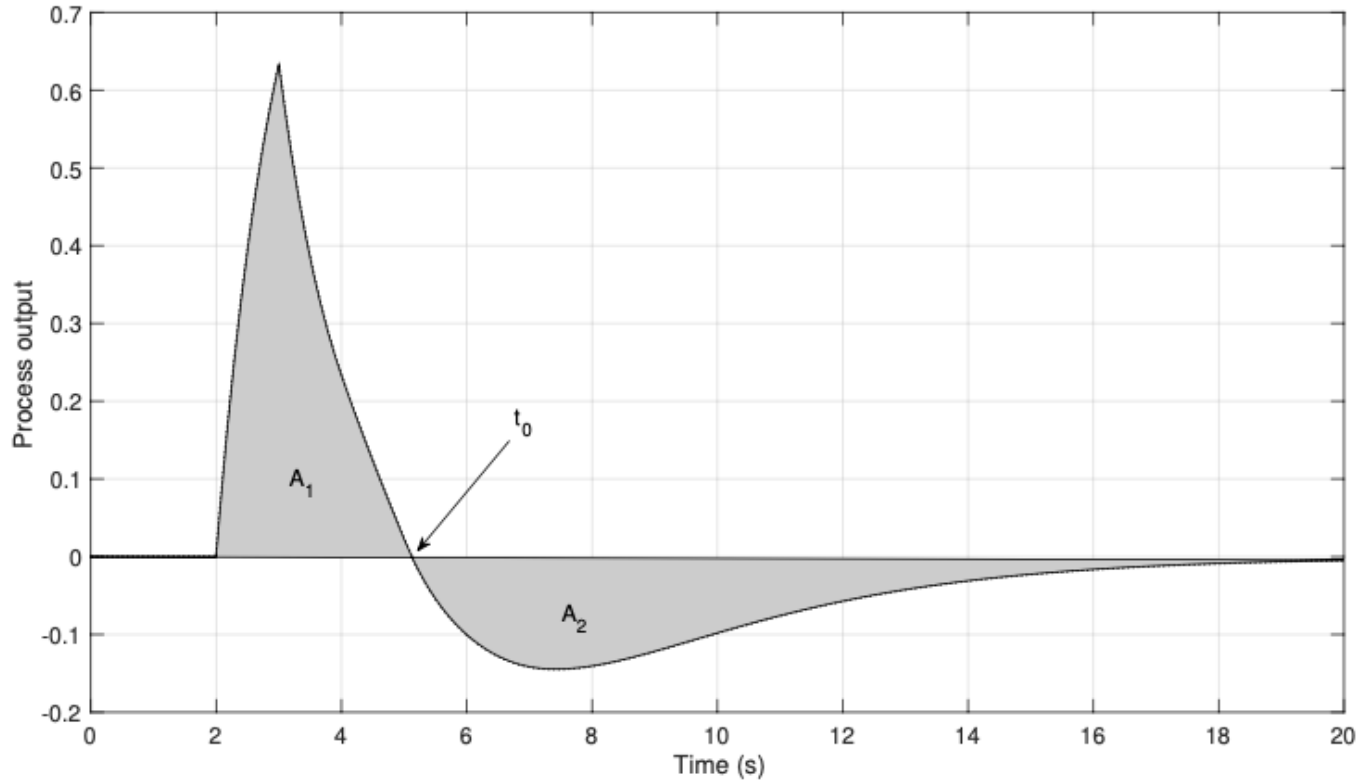
CS	Rule	M	$k_{ff}$	$T_p$	$T_z$
C	1	OS	$(K_d/K_u)e^{(L_d-L_u)/T_d}$	$T_d$	$T_u$
C	2	OS	$\frac{K_d(T_u+\lambda)}{K_u(T_d+\lambda)}\epsilon \quad T_u \neq T_d$ $\frac{K_d}{K_u}e^{(L_d-L_u)/T_d} \quad T_u = T_d$ $\epsilon = e^{(L_u/(T_u+\lambda)-L_d/(T_d+\lambda))}$	-	-
C	3	OS	$K_d/K_u - (K/\tau_i)IE$	$T_d$	$T_u$
C	4	OS	$K_d/K_u - (K/\tau_i)IE$	$T_d - (L_u - L_d)/4$	$T_u$
C	5	IAE	$(K_d(T_d + L_d))/(K_u(T_d + L_u))$	$T_d$	$T_u$
C	6	IAE	$\frac{K_d(T_u+\lambda)}{K_u(T_d+\lambda)}\theta \quad T_u \neq T_d$ $\frac{K_d(T_d+L_d)}{K_u(T_d+L_u)} \quad T_u = T_d$ $\theta = e^{-(L_u-L_d+(T_u-T_d)\log(2))/(T_d+\lambda)}$	-	-
C	7	IAE	$K_d/K_u - (K/\tau_i)IE$	$T_d - (L_u - L_d)/1.7$	$T_u$
C	8	ISE	$(K_d/K_u)e^{-(L_u-L_d)/(\lambda+T_d)}$	$T_d$	$T_u$
C	9	ISE	$K_d/K_u - (K/\tau_i)IE$	$T_d - (L_u - L_d)/\alpha$ $\alpha = \frac{L}{2T_d(1-e^{-L/(2T_d)})}$	$T_u$
C	10	IAE ISE OS	$K_d/K_u - (K/\tau_i)IE$	$T_d - \frac{(L_u - L_d)/\alpha}{\alpha}$ $\alpha = \begin{cases} \frac{L}{2T_d(1-e^{-L/(2T_d)})} & \text{ISE} \\ 1.7 & \text{IAE} \\ 4 & \text{OS} \end{cases}$	$T_u$
B	11	ISE	$K_d/K_u$	$\frac{3a-1-b+(a-1)\sqrt{1+4b}}{b-2} T_d$ 0 $a = T_u/T_d$ $b = a(a+1)e^{L/T_d}$ $b < 4a^2 - 2a$ or $b < a + \sqrt{a}$ otherwise	$\eta = \left(1 - \frac{2T_u}{b(T_d+T_p)}\right)$
B	12	IAE ISE	$K_d/K_u$	$T_d - (L_u - L_d)/\alpha$ $\alpha = \begin{cases} \frac{L}{2T_d(1-e^{-L/(2T_d)})} & \text{ISE} \\ 1.7 & \text{IAE} \end{cases}$	$T_u$

$IE = K_d(L_u - L_d + T_u - T_d + T_p - T_z)$

Static  $C_{ff} = k_{ff}$

Lead - lag  $C_{ff} = k_{ff} \frac{1 + sT_z}{1 + sT_p}$

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$$IE = -(\lambda + L_u)(K_d - k_{ff}K_u)$$

$$IAE = 2|A_1|$$

$$IAE = \int_0^{\infty} |e(t)|dt = \int_0^{t_0} e(t)dt - \int_{t_0}^{\infty} e(t)dt$$

$$IAE = 2 \int_0^{t_0} e(t)dt - IE_{+\infty}$$

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$$P_{ff} = \frac{K_d e^{-sL_d}}{1 + sT_d} - \frac{K_u e^{-sL_u}}{1 + sT_u} C_{ff}$$

$$E(s) = G_{e/d} \frac{1}{s} = -(\lambda + L_u) \left( \frac{sK_d e^{-sL_d}}{(sT_d + 1)(s\lambda + 1)} - \frac{sC_{ff}K_u e^{-sL_u}}{(sT_u + 1)(s\lambda + 1)} \right) \frac{1}{s}$$

$$G_{ol} = \frac{Y}{D} = \underbrace{P_d - P_u C_{ff}}_{P_{ff}}$$

$$S = \frac{1}{1 + K \frac{1 + s\tau_i}{s\tau_i} K_u \frac{1 - sL_u}{sT_u + 1}} = \frac{(\lambda + L_u)s}{\lambda s + 1}$$

$$y_{ol}(t) = \begin{cases} 0 & : 0 \leq t < L_d \\ K_d (1 - e^{(L_d-t)/T_d}) d & : L_d \leq t < L_u \\ \mathcal{L}^{-1}\{P_{ff}D\} & : t \geq L_u \end{cases}$$

$$y_{e/d}(t) = \begin{cases} 0 & : 0 \leq t < L_d \\ -\frac{K_d(\lambda + L_u) (e^{(L_d-t)/T_d}/T_d - e^{(L_d-t)/\lambda}/\lambda)}{\lambda - T_d} d & : L_d \leq t < L_u \\ \mathcal{L}^{-1}\{P_{ff}SD\} & : t \geq L_u \end{cases}$$

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## Classical scheme

**Rule 7** *Tuning rule for the classical control scheme to minimize IAE with a lead-lag compensator:*

1. Set  $T_z = T_u$  and  $L_{ff} = \max(0, L_d - L_u)$ .
2. Calculate  $T_p$  as:

$$T_p = \begin{cases} T_d & L_u - L_d \leq 0 \\ T_d - \frac{L_u - L_d}{1.7} & 0 < L_u - L_d < 1.7 T_d \\ 0 & L_u - L_d > 1.7 T_d \end{cases}$$

3. Calculate the compensator gain  $k_{ff}$  as:

$$k_{ff} = \frac{K_d}{K_u} - \frac{K}{\tau_i} IE$$

$$IE = \begin{cases} K_d (T_u - T_d + T_p - T_z) & L_d \geq L_u \\ K_d (L_u - L_d + T_u - T_d + T_p - T_z) & L_d < L_u \end{cases}$$

4. End of design.

## Non-interactive scheme

**Rule 12** *Tuning rule for the non-interacting control scheme to minimize ISE, IAE, or to remove the overshoot with a lead-lag compensator.*

1. Set  $k_{ff} = K_d/K_u$ ,  $T_z = T_u$  and  $L_{ff} = \max(0, L_d - L_u)$ .
2. Calculate  $L = L_u - L_d$ .
3. Calculate  $\alpha$  depending on the desired behaviour:

$$\alpha = \begin{cases} \frac{L}{2T_d(1-e^{-L/(2T_d)})} & \text{aggressive (ISE minimization)} \\ 1.7 & \text{moderate (IAE minimization)} \\ 4 & \text{conservative (overshoot removal)} \end{cases}$$

4. Set  $T_p$  according to:

$$T_p = \begin{cases} T_d & L \leq 0 \\ T_d - \frac{L}{\alpha} & 0 < L < \alpha T_d \\ 0 & L \geq \alpha T_d \end{cases}$$

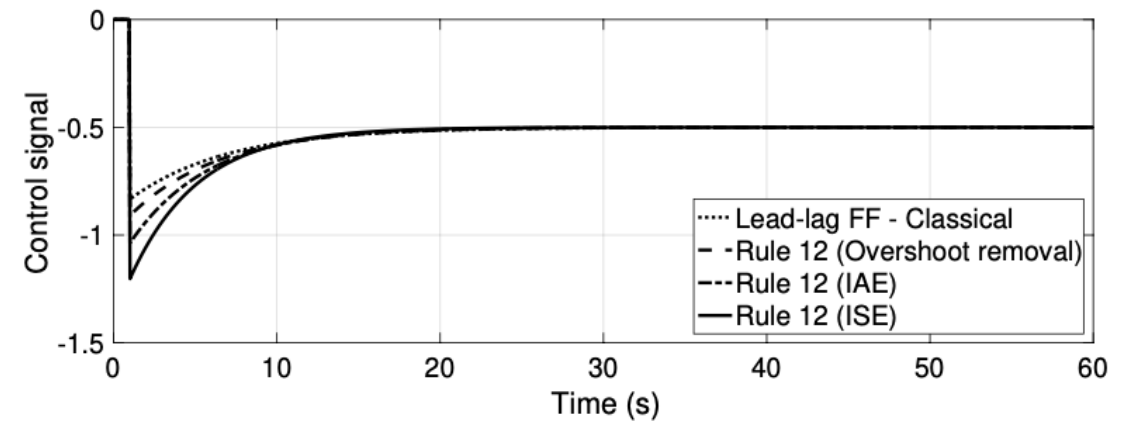
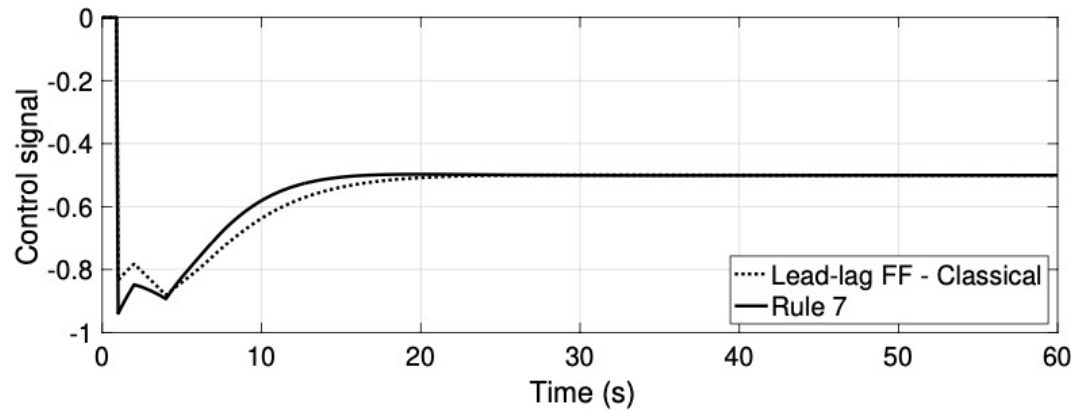
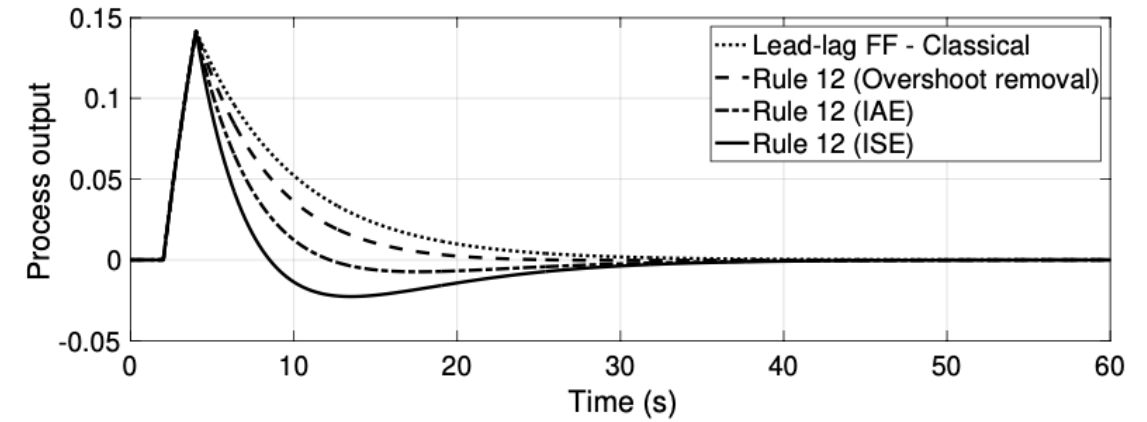
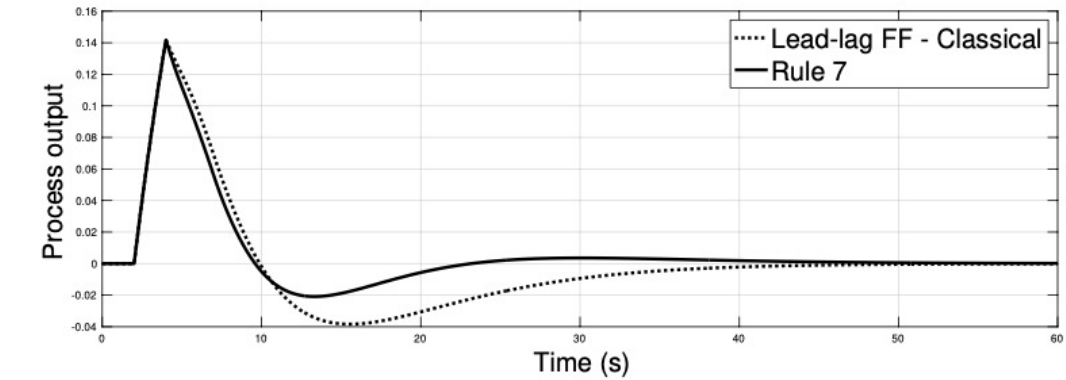
If  $T_p = 0$ , select a value close to zero to obtain a realizable compensator.

5. Set  $H(s)$  with Equation (4.66) for the non-interacting scheme.
6. End of design.

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## Classical scheme

## Non-interactive scheme



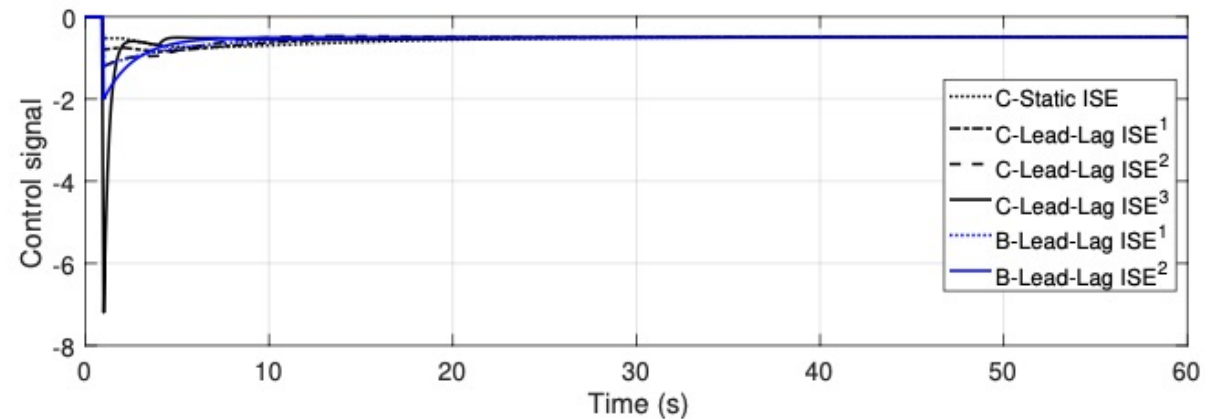
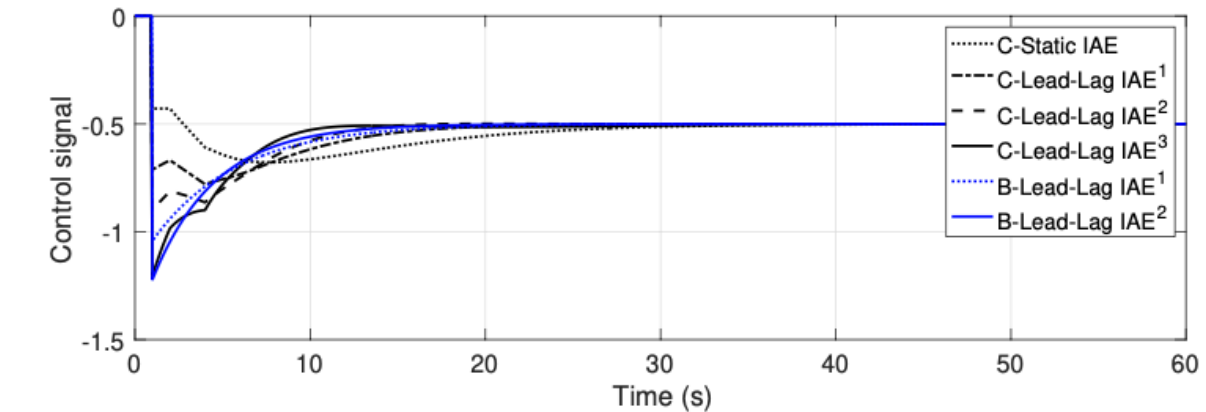
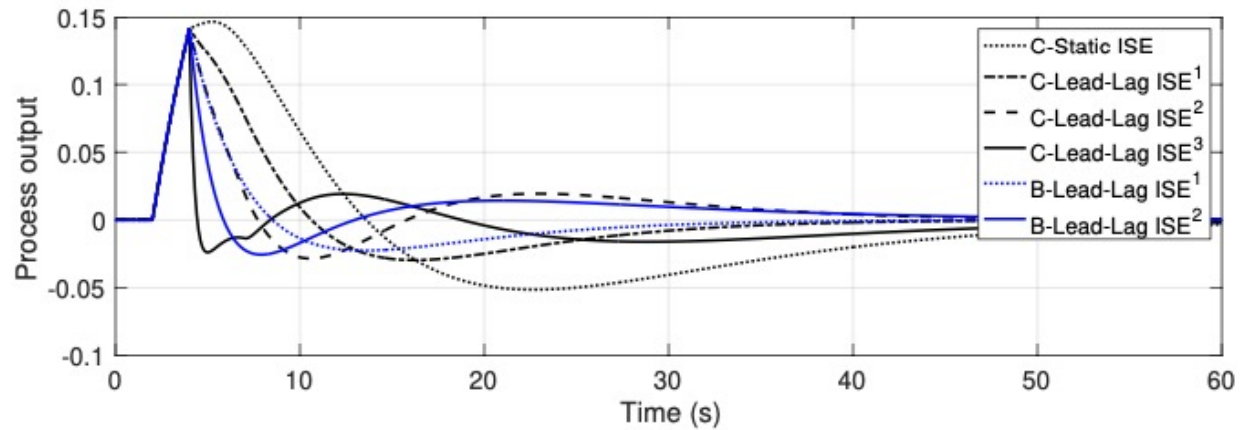
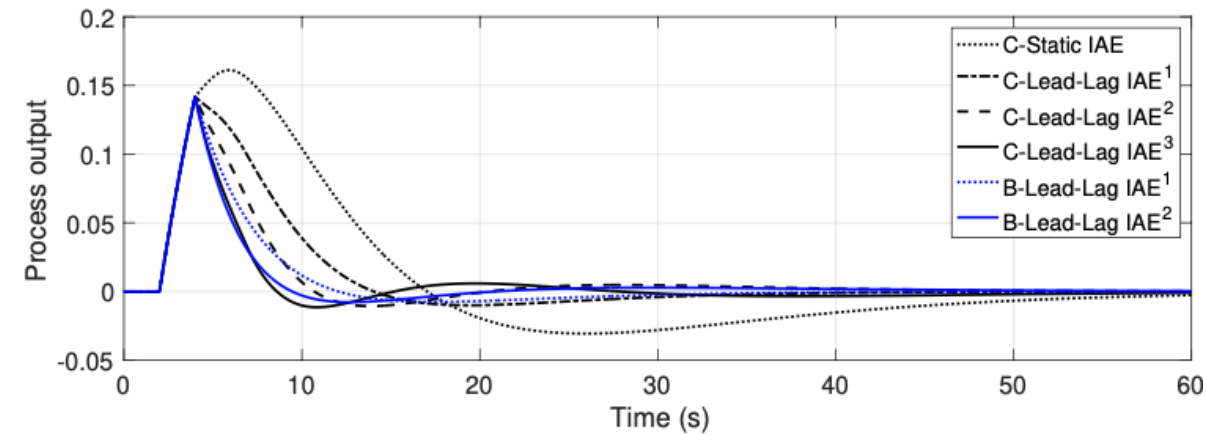
35% IAE reduction  
45% Overshoot reduction

38% IAE reduction  
47% ISE reduction

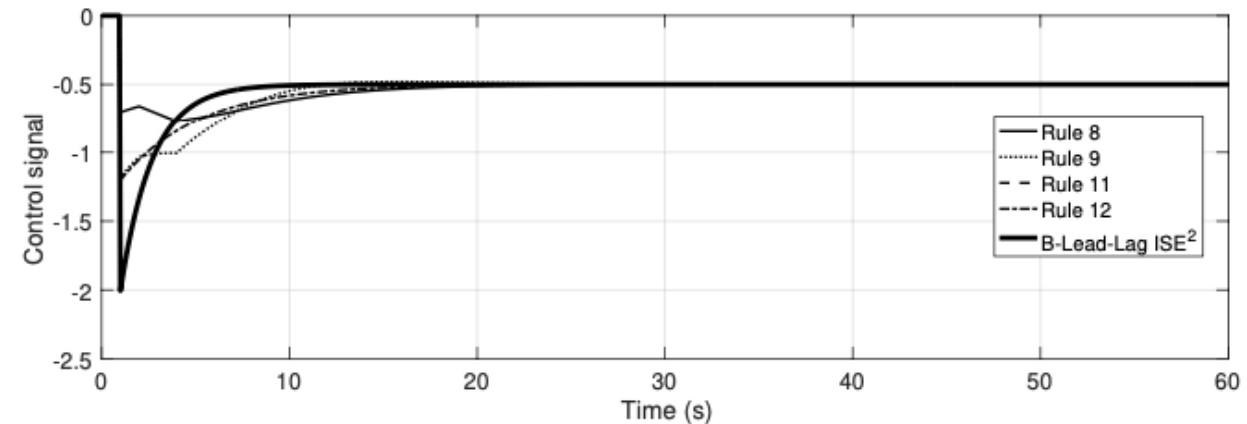
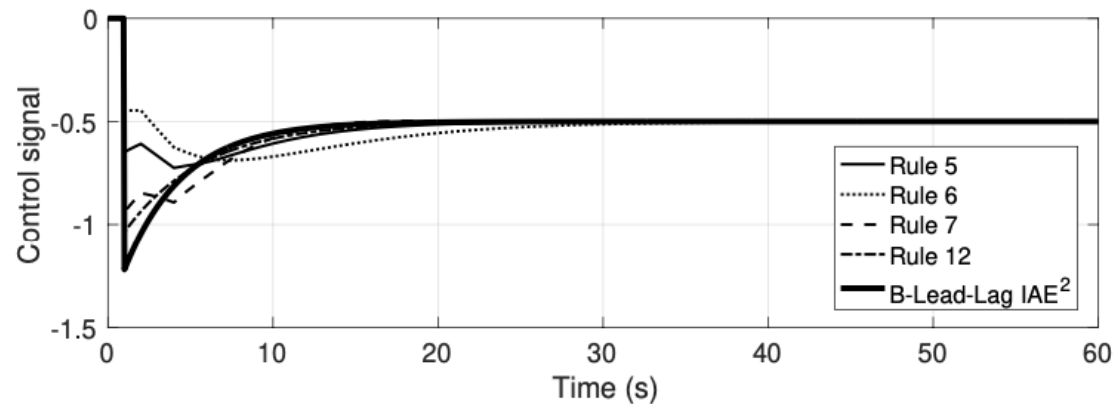
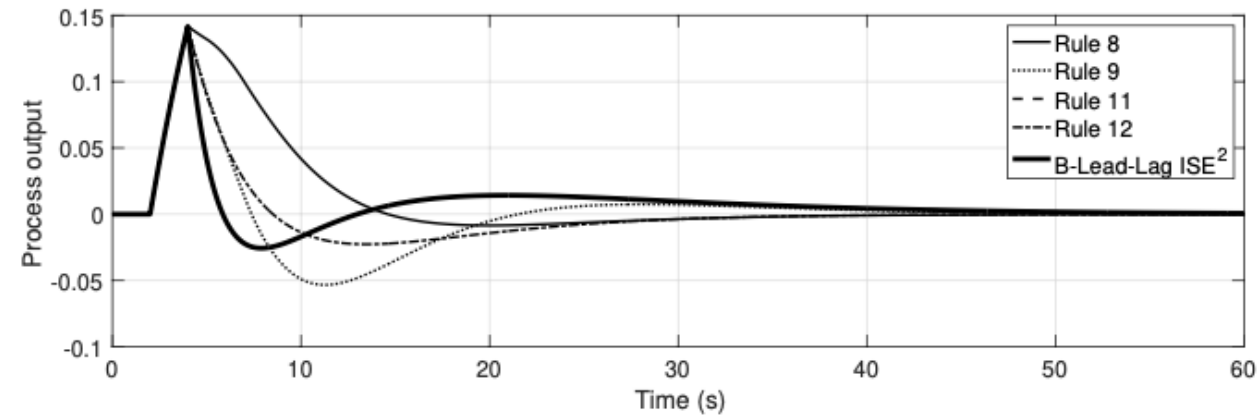
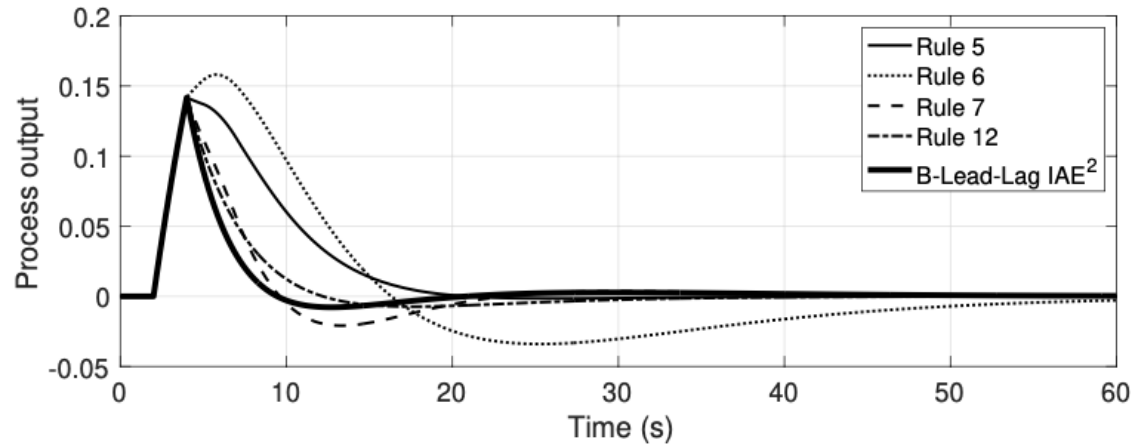
# 3

Selecting control schemes and tuning rules

## Obtaining optimal tuning values



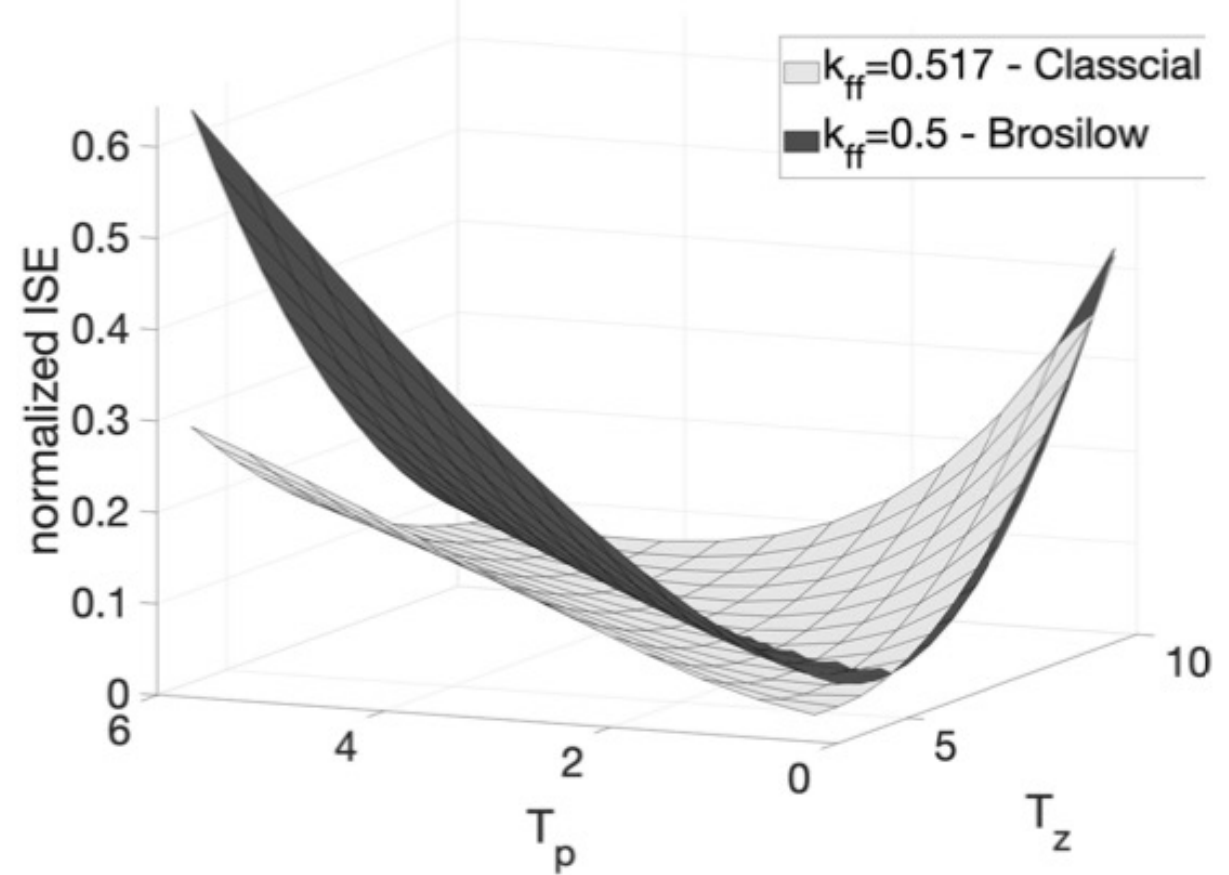
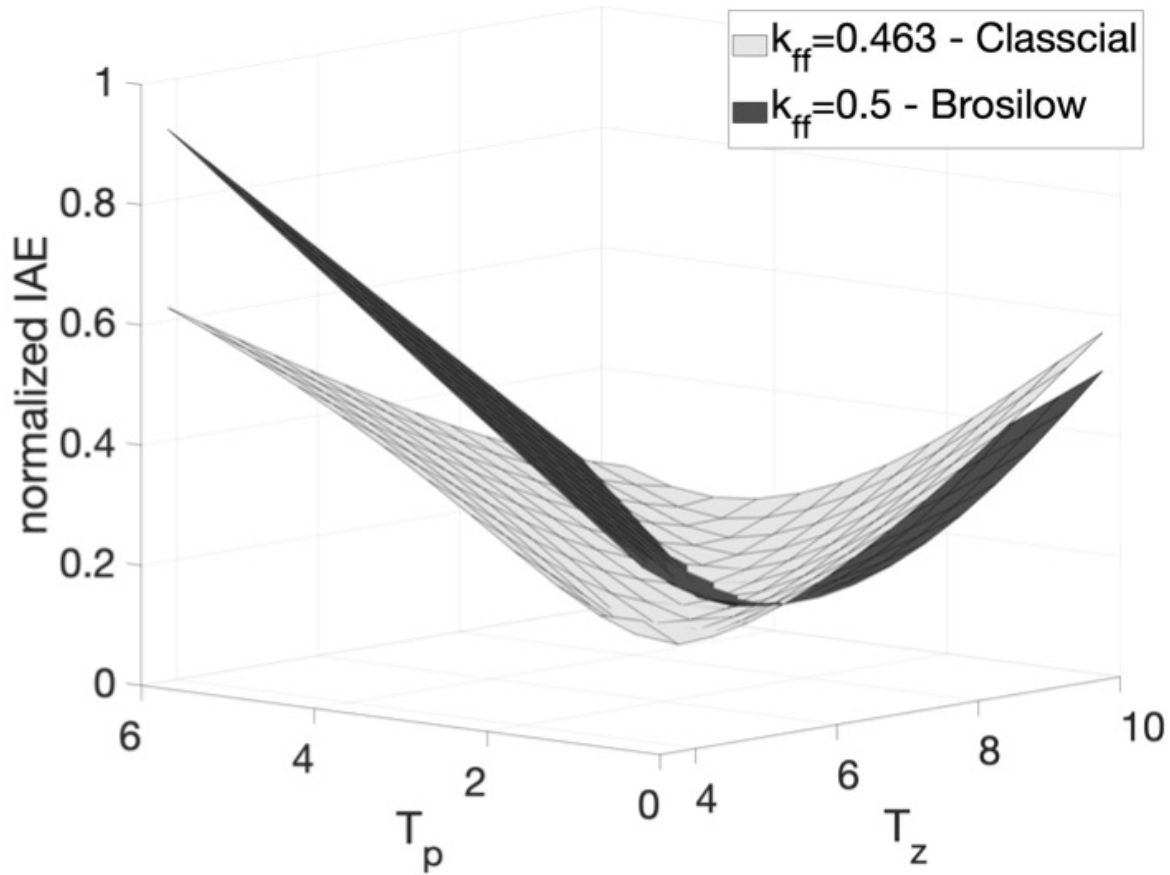
## Comparing the rules with the optimal rule



# Selecting control scheme and tuning rules

Rules	$IAE_{norm}$	$ISE_{norm}$	overshoot	peak	$k_{ff}$	$T_z$	$T_p$
FB	1	1	0	9.57	–	–	–
Static	0.477	0.182	9.1	44.69	0.5	–	–
Lead-Lag	0.257	0.082	7.67	76.28	0.5	10	6
Rule 1	0.284	0.139	0	36.47	0.358	10	6
Rule 2	0.514	0.324	2.5	23.54	0.306	–	–
Rule 3	0.236	0.116	0.22	44.15	0.386	10	6
Rule 4	0.18	0.084	0.56	57.97	0.414	10	5.5
Rule 5	0.232	0.113	0.3	45.07	0.389	10	6
Rule 6	0.461	0.199	6.8	37.6	0.446	–	–
Rule 7	0.166	0.059	4.18	87.79	0.453	10	4.824
Rule 8	0.212	0.094	1.7	54.29	0.422	10	6
Rule 9	0.207	0.061	10.66	136.17	0.491	10	4.158
Rule 10 (OS)	0.18	0.084	0.56	57.97	0.414	10	5.5
Rule 10 (IAE)	0.166	0.059	4.18	87.79	0.453	10	4.824
Rule 10 (ISE)	0.207	0.061	10.66	136.17	0.491	10	4.158
Static	0.682	0.363	0	0	0.5	–	–
Lead-Lag	0.229	0.089	0	66.67	0.5	10	6
Rule 11	0.151	0.033	5.13	301.57	0.5	6.789	1.691
Rule 12 (OS)	0.174	0.07	0.06	81.82	0.5	10	5.5
Rule 12 (IAE)	0.143	0.053	1.46	107.32	0.5	10	4.824
Rule 12 (ISE)	0.16	0.047	4.53	140.51	0.5	10	4.158

# Selecting control scheme and tuning rules

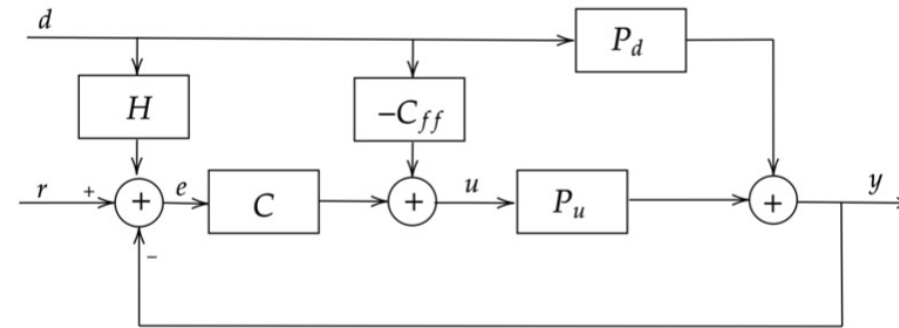
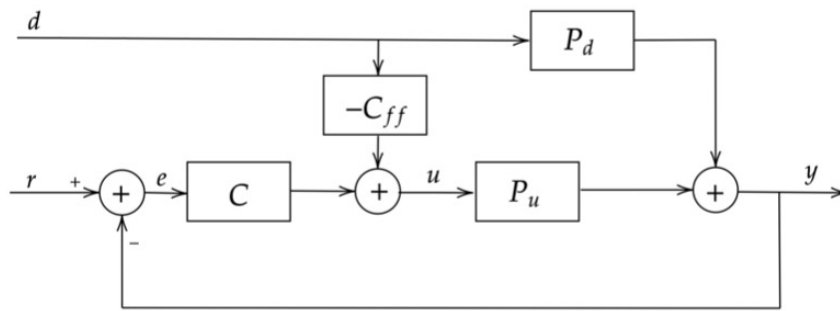




# 4

## Performance indices





$$I_{FF/FB} = 1 - \frac{IAE_{FF}}{IAE_{FB}}$$

where  $IAE_{FB}$  is the integrated absolute value of the control error obtained when only feedback is used, and  $IAE_{FF}$  is the corresponding  $IAE$  value obtained when feedforward is added to the loop.

As long as the feedforward improves control, i.e.  $IAE_{FF} < IAE_{FB}$ , the index is in the region  $0 < I_{FF/FB} < 1$ .

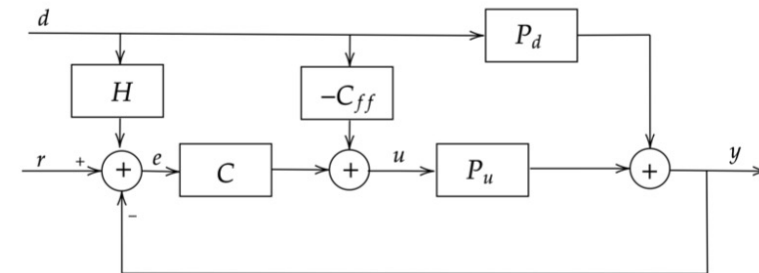
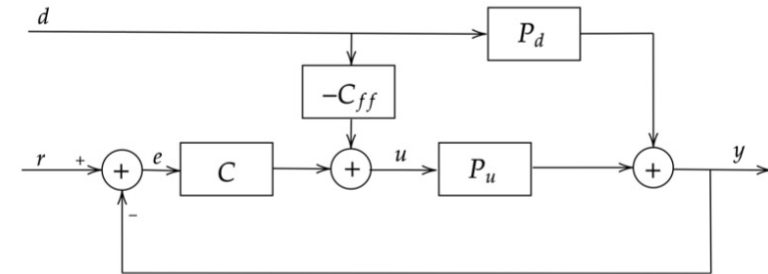
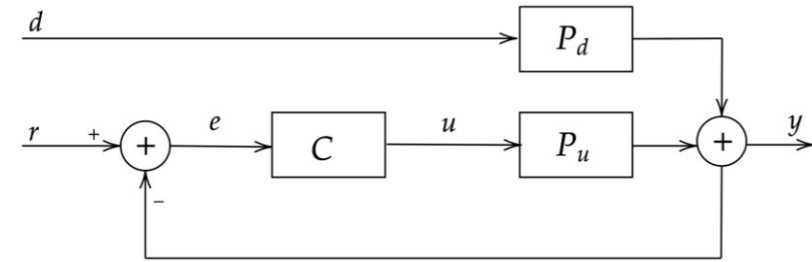
$$IAE_{FB} = K_d(\lambda + L_u)d$$

$$IAE_{FF} = K_d \left( L_1 - \tau \left( 1 - 2e^{-L_1/\tau} \right) \right) d$$

$$\tau = T_d - T_p \text{ and } L_1 = \max(0, L_u - L_d)$$

$$IAE_{FF} = dK_d(L_u - L_d)$$

$$\alpha = 1 - \frac{1}{a} + \frac{2}{a}e^{-a}, \quad a = \frac{L_u - L_d}{T_d - T_u - T_p + T_z}$$



Rules	$IAE^r$	$IAE^e$	$I_{FF/FB}^r$	$I_{FF/FB}^e$
FB	4.371	4.372	0	0
Static	2.116	1.549	0.516	0.646
Lead-lag	1.125	1.096	0.743	0.749
Rule 5	1.014	0.971	0.768	0.778
Rule 6	2.043	1.46	0.533	0.666
→ Rule 7	0.727	0.627	0.834	0.857
Lead-lag	1.0000	1.000	0.771	0.771
→ Rule 12	0.627	0.627	0.857	0.857

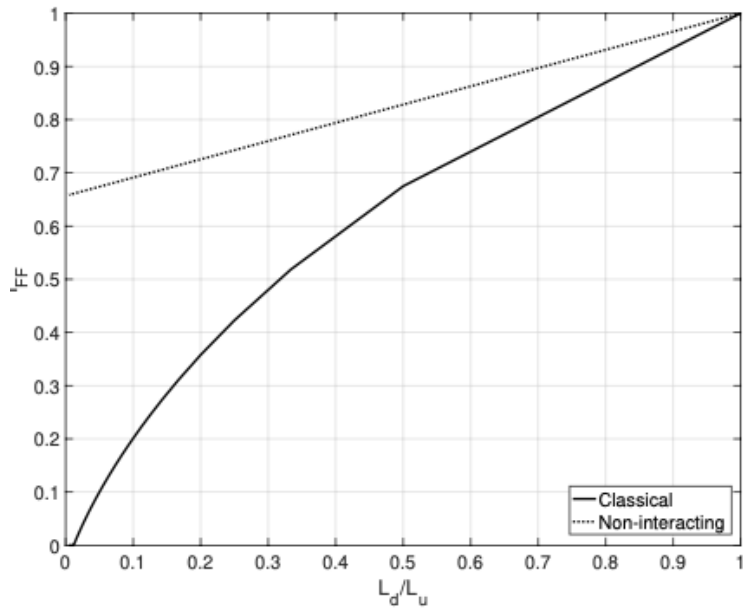
$$\frac{IAE_{\text{classical}}}{IAE_{\text{noninteracting}}} = \frac{2(L_u + \lambda)f(\lambda/T_d)}{T_d} \quad f(\lambda/T_d) = \begin{cases} \left(\frac{\lambda}{T_d}\right)^{-\frac{\lambda}{\lambda - T_d}} & \lambda \neq T_d \\ e^{-1} & \lambda = T_d \end{cases}$$

Therefore, the classical scheme gives a smaller IAE value when

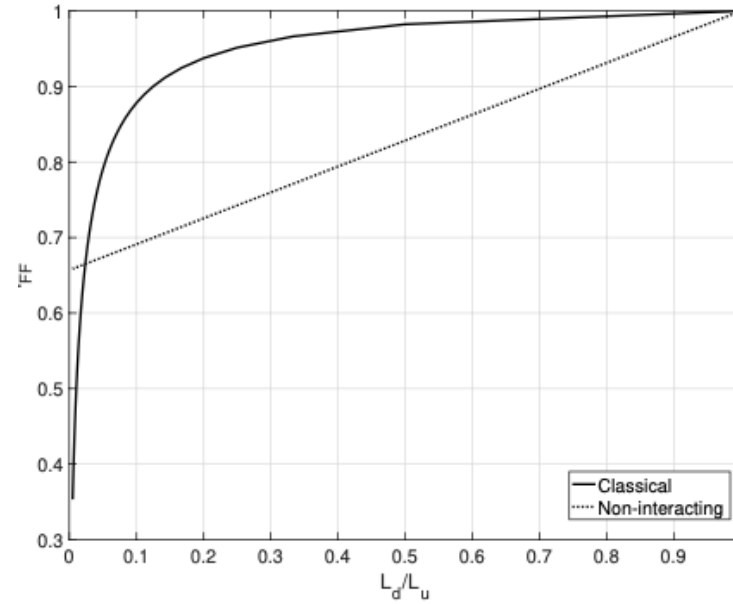
$$T_d > 2(L_u + \lambda)f(\lambda/T_d)$$

Since  $0 < f(\lambda/\tau) \leq 1$ , one can conclude that the classical scheme gives a better performance when  $T_d$  is large compared to process dead time  $L_u$  or the desired closed-loop time constant  $\lambda$ , i.e. when the load disturbance is varying slowly.

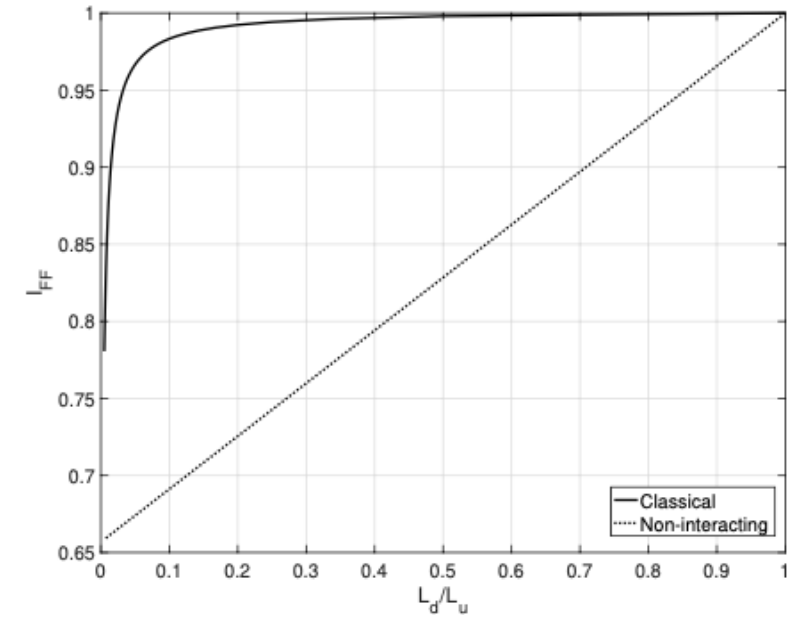
$$P_u = \frac{1}{1 + 10s} e^{-L_u s}, \quad P_d = \frac{1}{1 + T_d s} e^{-0.1s}$$



$T_d = 0.1$

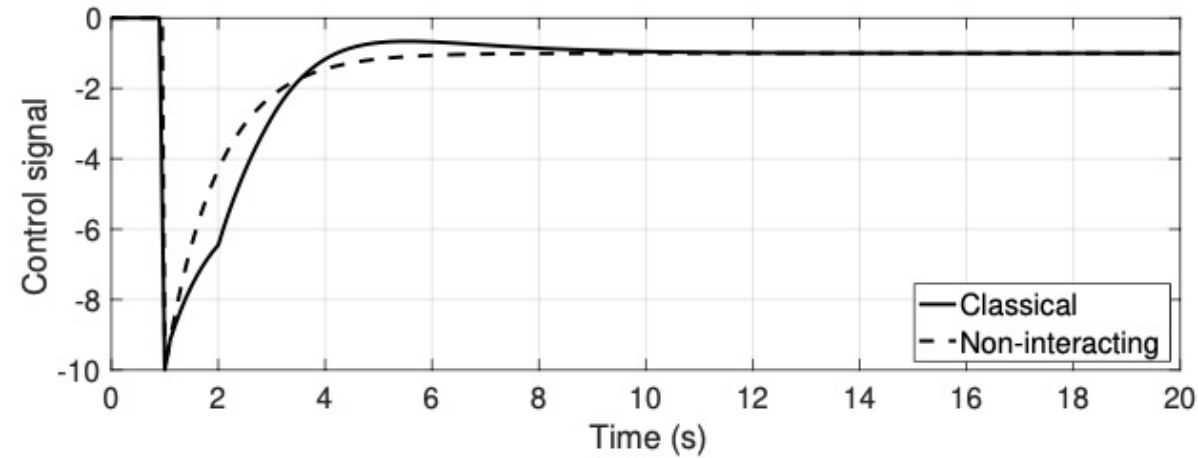
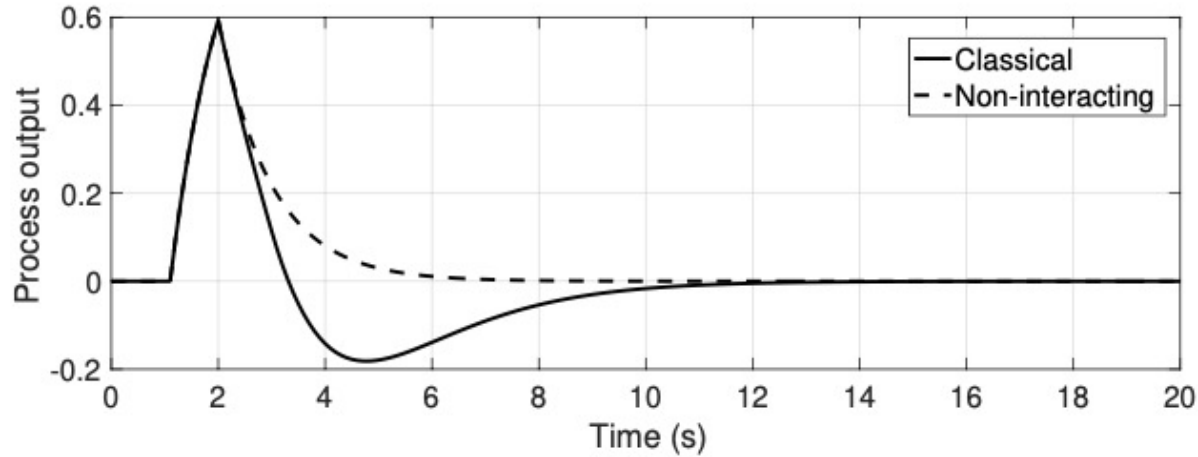


$T_d = 10$

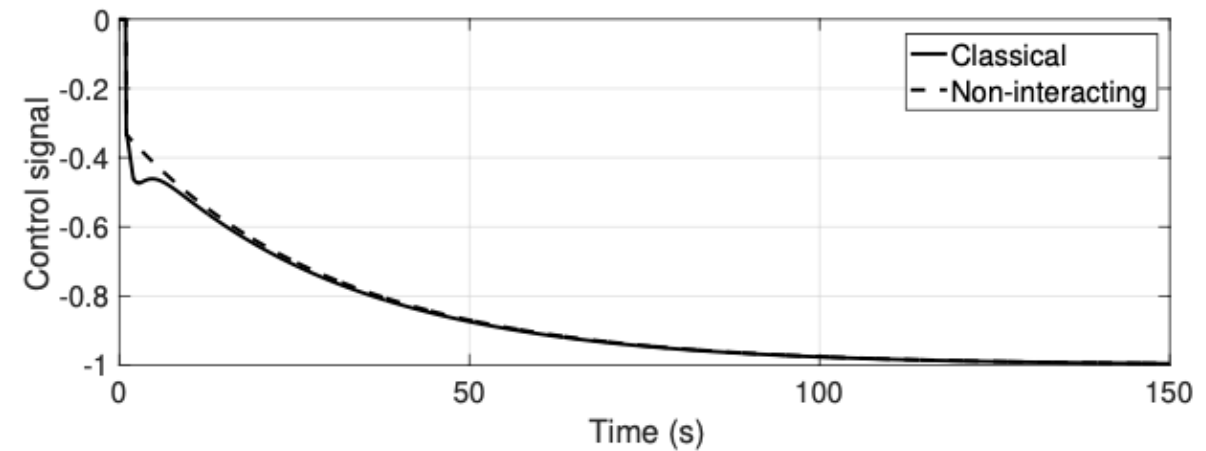
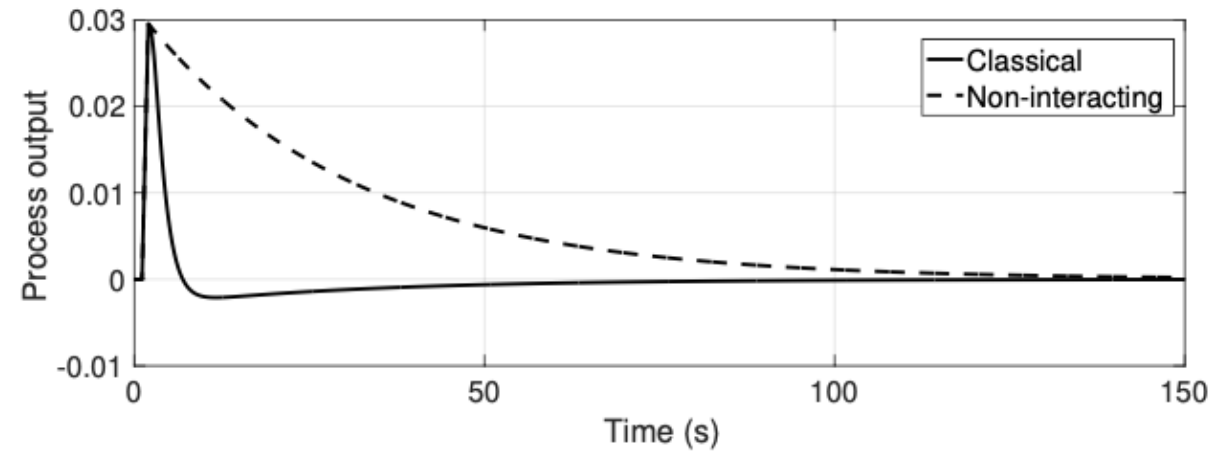


$T_d = 100$

$$T_d = 1 < 2(L_u + \lambda)f(\lambda/T_d) = 1.497$$



$$T_d = 30 > 2(L_u + \lambda)f(\lambda/T_d) = 4.832$$



# 5 Conclusions



- The motivation for feedforward tuning rules was introduced.
- The two available feedforward control schemes were used.
- Simple tuning rules based on the process and feedback controllers' parameters were presented for both control schemes.
- The proposed rules were compared with optimal tuning parameters.
- The effect of feedforward compensator parameters was analyzed and combined with the selection of the feedforward control schemes.
- Performance indices for feedforward control were proposed.

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